

Optical Waveguide Theory (G)



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Paderborn University — Summer Semester 2020

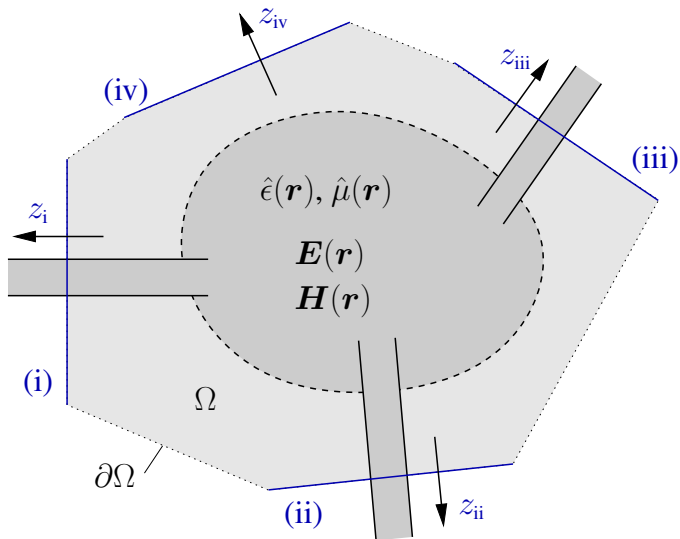
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Optical waveguide theory

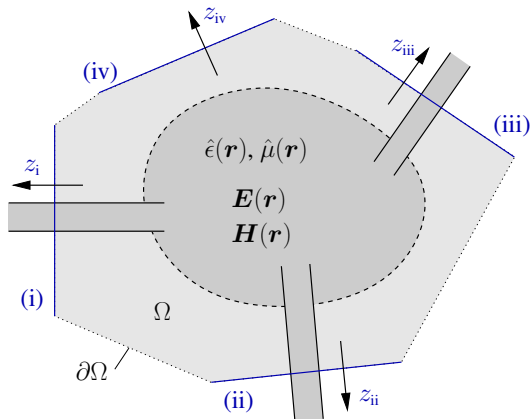
- A Photonics / integrated optics; theory, motto; phenomena, introductory examples.
- B Brush up on mathematical tools.
- C Maxwell equations, different formulations, interfaces, energy and power flow.
- D Classes of simulation tasks: scattering problems, mode analysis, resonance problems.
- E Normal modes of dielectric optical waveguides, mode interference.
- F Examples for dielectric optical waveguides.
- G Waveguide discontinuities & circuits, scattering matrices, reciprocal circuits.
- H Bent optical waveguides; whispering gallery resonances; circular microresonators.
- I Coupled mode theory, perturbation theory.
 - Hybrid analytical / numerical coupled mode theory.
- J A touch of photonic crystals; a touch of plasmonics.
 - Oblique semi-guided waves: 2-D integrated optics.
 - Summary, concluding remarks.

PICs, OICs, scattering matrices



Scattering matrices, prerequisites

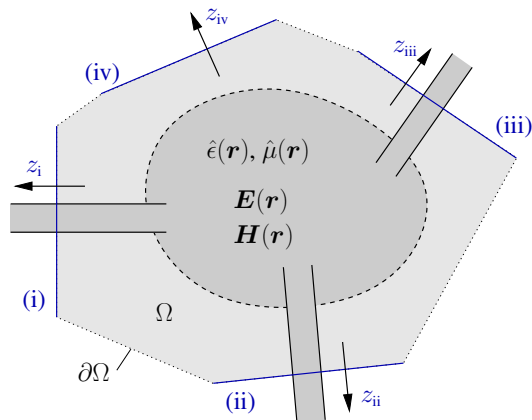
$\sim \exp(i\omega t)$ (FD)



- Passive, linear circuit.

Scattering matrices, prerequisites

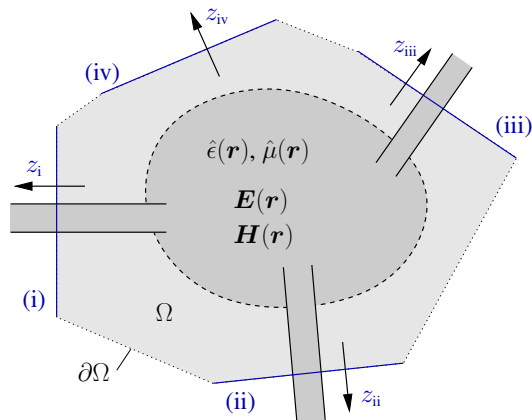
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- (Computational) domain of interest Ω , its boundary $\partial\Omega$.

Scattering matrices, prerequisites

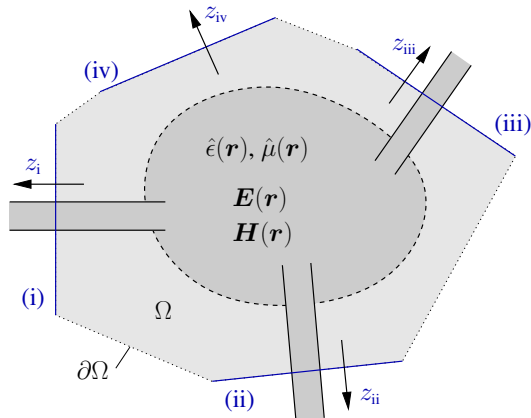
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- Connecting channels: lossless waveguides (or “half-spaces”).

Scattering matrices, prerequisites

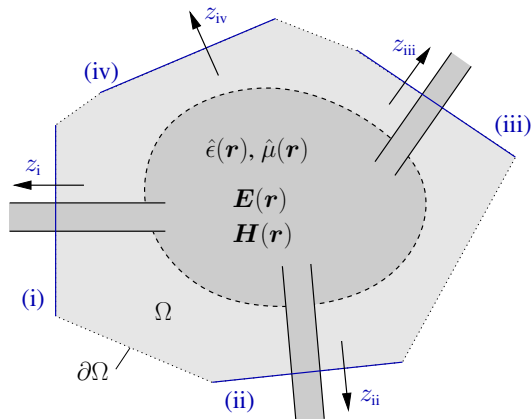
$\sim \exp(i\omega t)$ (FD)



- Physical **ports** $p = i, ii, \dots$: waveguide cross-section planes, local coordinates x_p, y_p, z_p ; local axis z_p oriented outwards of Ω .

Scattering matrices, prerequisites

$\sim \exp(i\omega t)$ (FD)

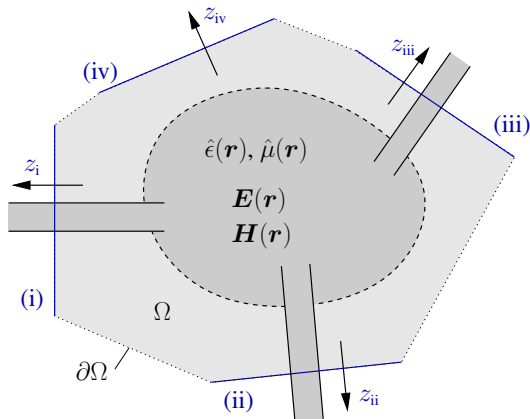


- Establish sets \mathcal{N}_p of *propagating* directional normal modes $\{\psi_{p,m}^d := (\mathbf{E}_{p,m}^d, \mathbf{H}_{p,m}^d), \beta_{p,m}; d = \text{f,b}\}$ on each port p .

(Restriction to propagating fields: a condition on port positioning / a model assumption.)

Scattering matrices, prerequisites

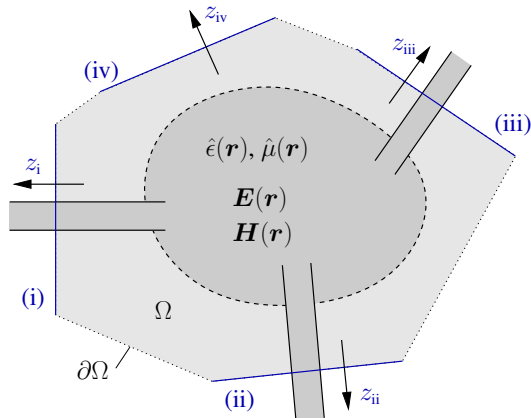
$\sim \exp(i\omega t)$ (FD)



- Ports & modes are such that all mode fields vanish
 - on all “other” port planes, and
 - on $\partial\Omega$ outside the ports.

Scattering matrices, prerequisites

$\sim \exp(i\omega t)$ (FD)



Field on port plane p and “outside”:

$$\begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix} (x_p, y_p, z_p) = \sum_{m \in \mathcal{N}_p} F_{p,m} \psi_{p,m}^f(x_p, y_p) e^{-i\beta_{p,m}z_p} + B_{p,m} \psi_{p,m}^b(x_p, y_p) e^{i\beta_{p,m}z_p}$$

Scattering matrices, prerequisites

$\sim \exp(i\omega t)$ (FD)

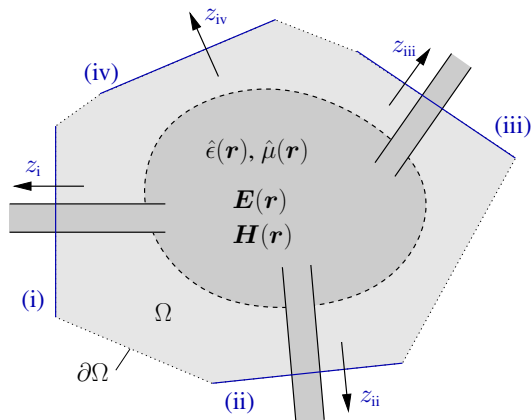
- Passive, linear circuit.
- (Computational) domain of interest Ω , its boundary $\partial\Omega$.
- Connecting channels: lossless waveguides (or “half-spaces”).
- Physical ports $p = \text{i, ii, } \dots$: waveguide cross-section planes, local coordinates x_p, y_p, z_p ; local axis z_p oriented outwards of Ω .
- Establish sets \mathcal{N}_p of *propagating* directional normal modes $\{\psi_{p,m}^d := (\mathbf{E}_{p,m}^d, \mathbf{H}_{p,m}^d), \beta_{p,m}; d = \text{f,b}\}$ on each port p .
(Restriction to propagating fields: a condition on port positioning / a model assumption.)
- Ports & modes are such that all mode fields vanish on all “other” port planes, and on $\partial\Omega$ outside the ports.

↪ Field on port plane p and “outside”:

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Scattering matrices

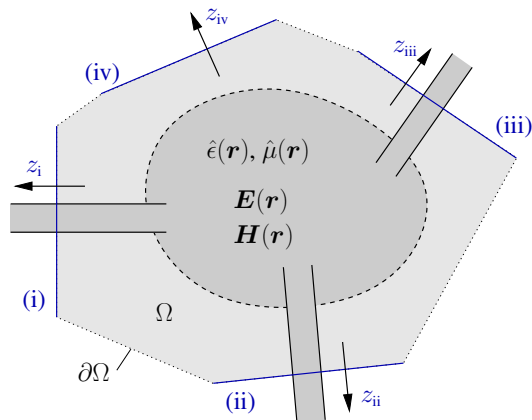
$\sim \exp(i\omega t)$ (FD)



- Merge all mode indices $\{m\}$ and port IDs $\{p\}$ into one set of mode identifiers $\{\nu\}$, $\mathcal{N} = \cup_p \mathcal{N}_p$.

Scattering matrices

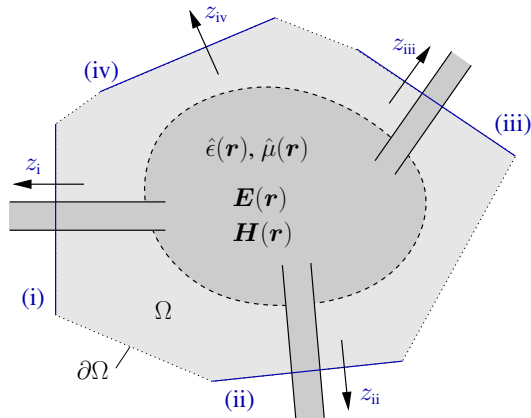
$\sim \exp(i\omega t)$ (FD)



- Assert that $\psi_{p,\cdot}(\mathbf{r}) = 0$ for all $\mathbf{r} \in \partial\Omega$, $\mathbf{r} \notin \text{port } p$.

Scattering matrices

$\sim \exp(i\omega t)$ (FD)

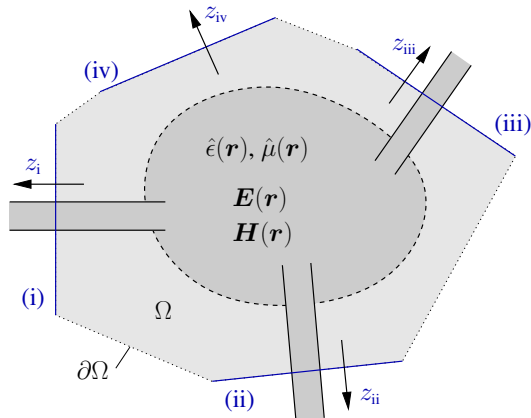


- Field on $\partial\Omega$:
$$\begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix} = \sum_{\nu \in \mathcal{N}} \{F_{\nu} \psi_{\nu}^f + B_{\nu} \psi_{\nu}^b\}.$$

(Position arguments omitted.)

Scattering matrices

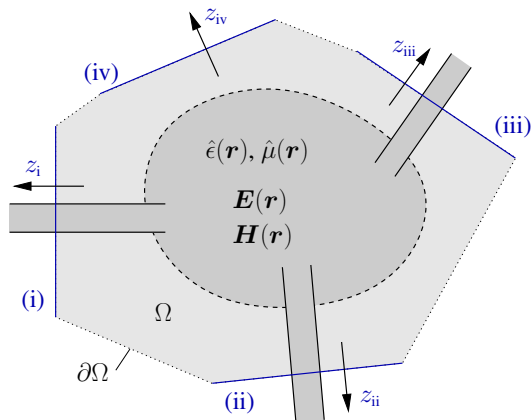
$\sim \exp(i\omega t)$ (FD)



- B_ν : \sim incident modes, traveling towards the interior of Ω .
- F_ν : \sim outgoing modes, traveling towards the exterior of Ω .
- Combine into amplitude vectors \mathbf{B}, \mathbf{F} .

Scattering matrices

$\sim \exp(i\omega t)$ (FD)

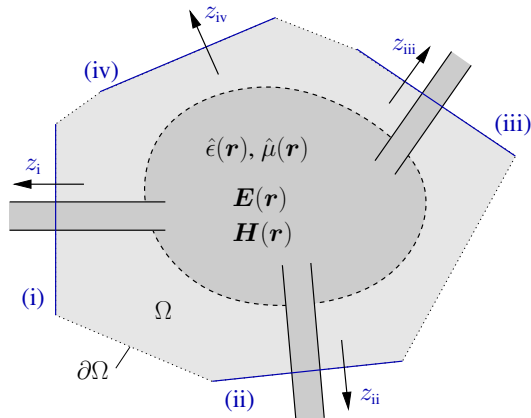


Linear circuit \longleftrightarrow linear dependence of \mathbf{F} on \mathbf{B} ,

Scattering matrix \mathbf{S} of the circuit: $\mathbf{F} = \mathbf{S}\mathbf{B}$, $\mathbf{S} = (S_{\nu\mu})$.

Scattering matrices

$\sim \exp(i\omega t)$ (FD)



- $S_{\nu\nu}$: $\sim (\nu, \mathbf{b}) \rightarrow (\nu, \mathbf{f})$, reflection coefficient for mode ν .
- $S_{\nu\mu}$: $\sim (\mu, \mathbf{b}) \rightarrow (\nu, \mathbf{f})$, transmission coefficient for modes μ, ν .

Scattering matrices

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 - Field on $\partial\Omega$:
$$\begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix} = \sum_{\nu \in \mathcal{N}} \{F_\nu \psi_\nu^f + B_\nu \psi_\nu^b\}.$$
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 - B_ν : \sim incident modes, traveling towards the interior of Ω .
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-

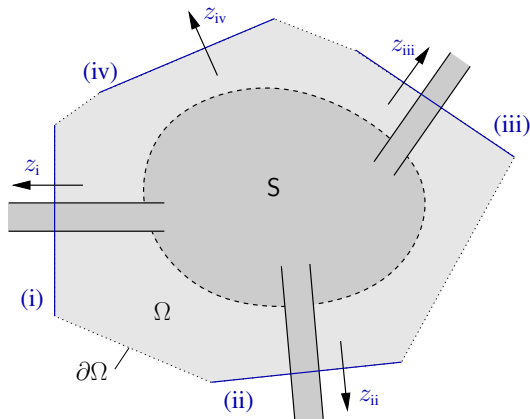
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PICs, OICs, scattering matrices, scenarios

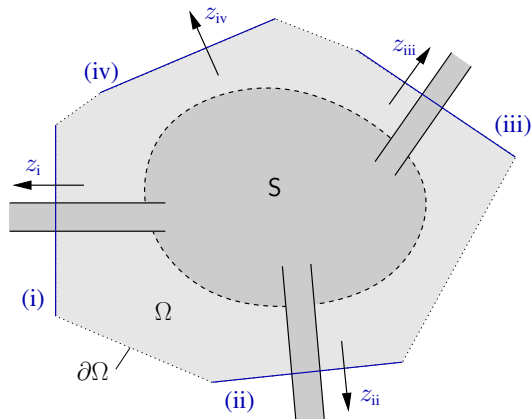
$\sim \exp(i\omega t)$ (FD)



- Scenario: Full matrix \mathbf{S} , including guided and radiation modes, large $\dim \mathbf{S} \leftrightarrow$ theoretical results.

PICs, OICs, scattering matrices, scenarios

$\sim \exp(i\omega t)$ (FD)

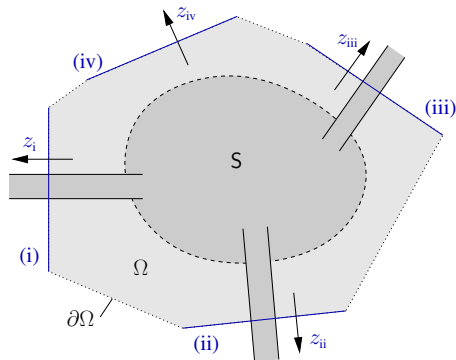


- Scenario: Restrict to a specific set of (guided) modes, or:
Only a small set of guided modes are relevant:
small $\dim \mathbf{S} = N \times N \leftrightarrow$ an N -port circuit, a $2N$ -pole.

(N : the total number of relevant modes, not the number of ports.)

Scattering matrices, port plane positions

$\sim \exp(i\omega t)$ (FD)



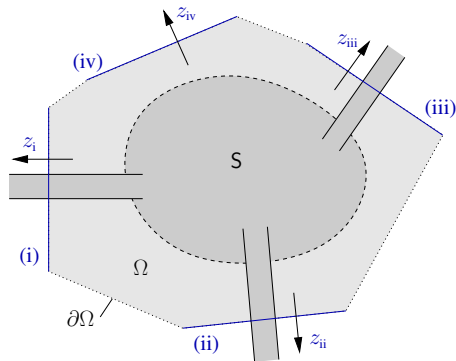
- Shift port plane of mode ν by Δz_ν : $F_\nu \rightarrow F'_\nu = F_\nu e^{-i\beta_\nu \Delta z_\nu}$,
Shift port plane of mode μ by Δz_μ : $B_\mu \rightarrow B'_\mu = B_\mu e^{i\beta_\mu \Delta z_\mu}$,
↪ $F'_\nu = S'_{\nu\mu} B'_\mu$, $S'_{\nu\mu} = S_{\nu\mu} e^{-i(\beta_\nu \Delta z_\nu + \beta_\mu \Delta z_\mu)}$.

(Moving port planes \leftrightarrow Phase change in reflection/transmission coefficients.)

(Moving port planes \leftrightarrow No effect on reflectances/transmittances.)

Scattering matrices, port mode orthogonality

$\sim \exp(i\omega t)$ (FD)



- Orthogonality relations on port plane p :

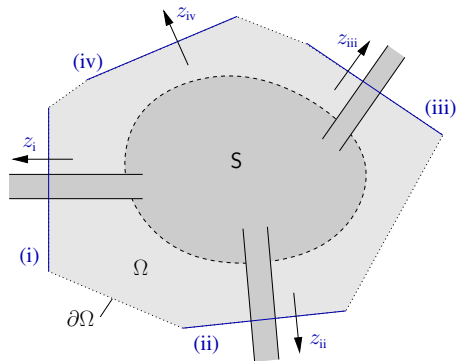
$$(\mathbf{E}_a, \mathbf{H}_a; \mathbf{E}_b, \mathbf{H}_b) = \frac{1}{4} \iint_p (E_{ax}^* H_{by} - E_{ay}^* H_{bx} + H_{ay}^* E_{bx} - H_{ax}^* E_{by}) dx_p dy_p$$

$$(\psi_{p,l}^d; \psi_{p,m}^r) = \pm \delta_{dr} \delta_{lm} P_{p,m}.$$

(Things restricted to propagating modes.)

Scattering matrices, port mode orthogonality

$\sim \exp(i\omega t)$ (FD)



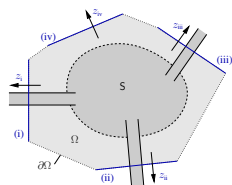
- Extend to the full boundary $\partial\Omega$:

$$(\mathbf{E}_a, \mathbf{H}_a; \mathbf{E}_b, \mathbf{H}_b) := \frac{1}{4} \int_{\partial\Omega} (\mathbf{E}_a^* \times \mathbf{H}_b + \mathbf{E}_b \times \mathbf{H}_a^*) \cdot d\mathbf{a}$$

$$\hookrightarrow (\psi_{p,l}^d; \psi_{q,m}^r) = \pm \delta_{dr} \delta_{pq} \delta_{lm} P_{p,m} \quad \text{or} \quad (\psi_\nu^d; \psi_\mu^r) = \pm \delta_{dr} \delta_{\nu\mu} P_\nu.$$

(Modes belonging to different ports are mutually orthogonal.)

Scattering matrices, power balance

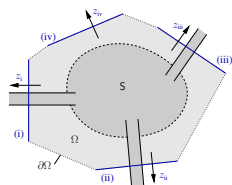


$\sim \exp(i\omega t)$ (FD)

- Net power outflow across the border of the circuit:

$$P = \int_{\partial\Omega} \mathbf{S} \cdot d\mathbf{a} = (\mathbf{E}, \mathbf{H}; \mathbf{E}, \mathbf{H}) = \sum_p \sum_{m \in \mathcal{N}_p} (|F_{p,m}|^2 - |B_{p,m}|^2) P_{p,m}$$

Scattering matrices, power balance

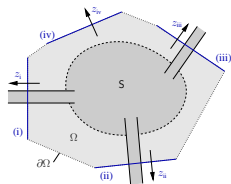


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Scattering matrices, power balance



$\sim \exp(i\omega t)$ (FD)

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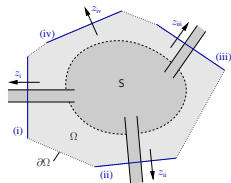
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$|B_\mu|^2 P_\mu$: incident power carried by mode μ ,

$|F_\nu|^2 P_\nu$: outgoing power carried by mode ν ,

$$F_\nu = S_{\nu\mu} B_\mu.$$

Scattering matrices, power balance



$\sim \exp(i\omega t)$ (FD)

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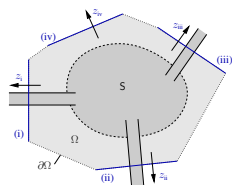
$|F_\nu|^2 P_\nu$: outgoing power carried by mode ν ,

$$F_\nu = S_{\nu\mu} B_\mu.$$

$$|S_{\nu\mu}|^2 \frac{P_\nu}{P_\mu} = \frac{|F_\nu|^2 P_\nu}{|B_\mu|^2 P_\mu}, \quad \begin{array}{l} \mu \neq \nu: \text{ power transmittance } \mu \rightarrow \nu, \\ \mu = \nu: \text{ power reflectance for mode } \nu. \end{array}$$

(Uniform normalized modes, $P_\nu = P_\mu$: transmittances are directly given by elements of the scattering matrix).

Scattering matrices, power balance



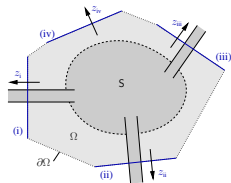
$\sim \exp(i\omega t)$ (FD)

- Net power outflow across the border of the circuit:

$$P = \int_{\partial\Omega} \mathbf{S} \cdot d\mathbf{a} = (\mathbf{E}, \mathbf{H}; \mathbf{E}, \mathbf{H}) = P_0 \left(\mathbf{B}^* \cdot (\mathbf{S}^\dagger \mathbf{S} - \mathbf{1}) \mathbf{B} \right),$$

uniform normalization, $P_\nu = P_0$ for all ν .

Scattering matrices, power balance



$\sim \exp(i\omega t)$ (FD)

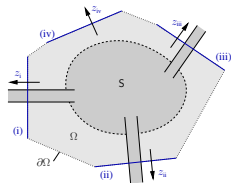
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uniform normalization, $P_\nu = P_0$ for all ν .

- Lossless circuit $\iff \int_{\partial\Omega} \mathbf{S} \cdot d\mathbf{a} = 0 \iff \mathbf{S}^\dagger \mathbf{S} = \mathbf{1}$,
the scattering matrix of a lossless circuit is unitary.

Scattering matrices, power balance



$\sim \exp(i\omega t)$ (FD)

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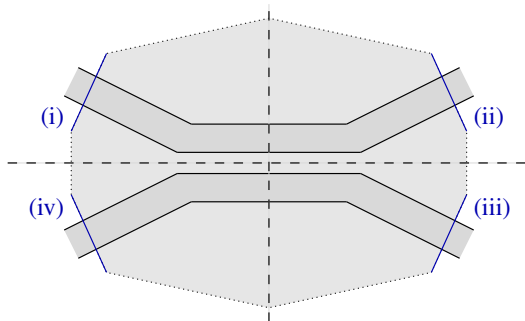
- Lossless circuit $\iff \int_{\partial\Omega} \mathbf{S} \cdot d\mathbf{a} = 0 \iff \mathbf{S}^\dagger \mathbf{S} = \mathbf{1}$,
the scattering matrix of a lossless circuit is unitary.

- Lossy circuit $\iff \int_{\partial\Omega} \mathbf{S} \cdot d\mathbf{a} \leq 0 \iff \mathbf{B}^* \cdot \mathbf{S}^\dagger \mathbf{S} \mathbf{B} \leq \mathbf{B}^* \mathbf{B}$,

$$\sum_{\nu} |S_{\nu\mu}|^2 \leq 1 \text{ for all } \mu. \quad \text{(The sum of transmittances mode } \mu \text{ to all other modes } \nu \text{ is less than one.)}$$

(Interior lossy media, or radiative losses: outgoing propagating modes not taken into account.)

Scattering matrices, symmetry

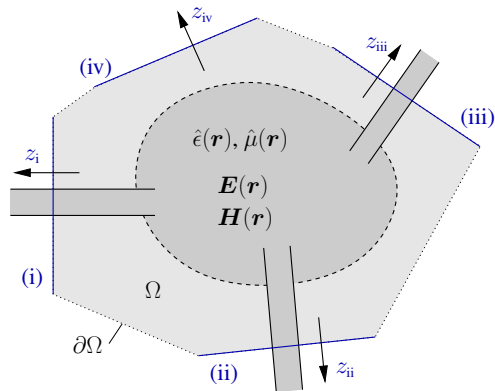


Circuit with specific spatial symmetry
& symmetrical setting of the port planes

↪ respective symmetry in related coefficients of \mathbf{S} ,
symmetric power transmission properties.

Scattering matrices, reciprocity

$\sim \exp(i\omega t)$ (FD)

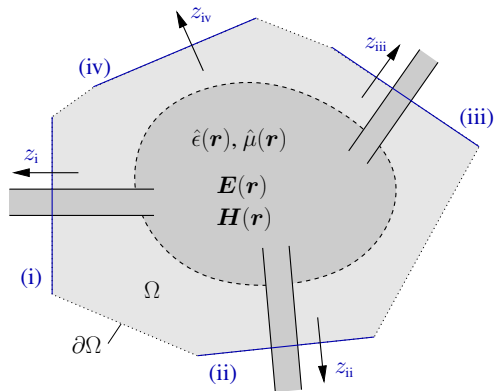


Circuit properties
for reversed
wave propagation ?

$$S_{\nu\mu} \longleftrightarrow S_{\mu\nu} ?$$

Scattering matrices, reciprocity

$\sim \exp(i\omega t)$ (FD)



Circuit properties
for reversed
wave propagation ?

$$S_{\nu\mu} \longleftrightarrow S_{\mu\nu} ?$$

- E_1, H_1 and E_2, H_2 solve $\nabla \times E = -i\omega\mu_0\hat{\mu}H$, $\nabla \times H = i\omega\epsilon_0\hat{\epsilon}E$.

$$\hookrightarrow \nabla \cdot (E_1 \times H_2 + H_1 \times E_2) = 0, \quad \text{if } \hat{\epsilon} \text{ and } \hat{\mu} \text{ are symmetric.}$$

(i.e. if $\hat{\epsilon}^T = \hat{\epsilon}$, $\hat{\mu}^T = \hat{\mu}$.)

(Note: order of factors, no complex conjugates.)

Scattering matrices, reciprocity

- $\mathbf{E}_1, \mathbf{H}_1$ and $\mathbf{E}_2, \mathbf{H}_2$ solve $\nabla \times \mathbf{E} = -i\omega\mu_0\hat{\mu}\mathbf{H}$, $\nabla \times \mathbf{H} = i\omega\epsilon_0\hat{\epsilon}\mathbf{E}$
 ↪ $\nabla \cdot (\mathbf{E}_1 \times \mathbf{H}_2 + \mathbf{H}_1 \times \mathbf{E}_2) = 0$, if $\hat{\epsilon}$ and $\hat{\mu}$ are symmetric,

Scattering matrices, reciprocity

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↳ $0 = \int_{\Omega} \nabla \cdot (\mathbf{E}_1 \times \mathbf{H}_2 + \mathbf{H}_1 \times \mathbf{E}_2) d^3r = \int_{\partial\Omega} (\mathbf{E}_1 \times \mathbf{H}_2 + \mathbf{H}_1 \times \mathbf{E}_2) \cdot d\mathbf{a}.$

Scattering matrices, reciprocity

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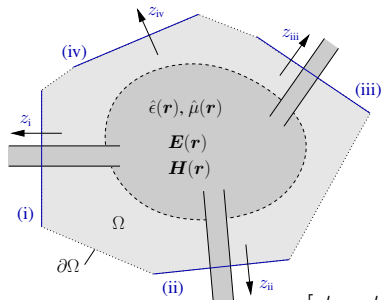
- Fields on $\partial\Omega$: $\begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix}_j = \sum_{\nu \in \mathcal{N}} \{F_{j,\nu}\psi_{\nu}^f + B_{j,\nu}\psi_{\nu}^b\}, \quad j = 1, 2,$

$$[\psi_a; \psi_b] := \int_{\partial\Omega} (\mathbf{E}_a \times \mathbf{H}_b + \mathbf{H}_a \times \mathbf{E}_b) \cdot d\mathbf{a},$$

↪ $0 = \sum_{\nu} \sum_{\mu} \left(F_{1,\nu}F_{2,\mu}[\psi_{\nu}^f; \psi_{\mu}^f] + F_{1,\nu}B_{2,\mu}[\psi_{\nu}^f; \psi_{\mu}^b] \right. \\ \left. + B_{1,\nu}F_{2,\mu}[\psi_{\nu}^b; \psi_{\mu}^f] + B_{1,\nu}B_{2,\mu}[\psi_{\nu}^b; \psi_{\mu}^b] \right).$

Scattering matrices, reciprocity

$\sim \exp(i\omega t)$ (FD)



$$[\psi_a; \psi_b] := \int_{\partial\Omega} (\mathbf{E}_a \times \mathbf{H}_b + \mathbf{H}_a \times \mathbf{E}_b) \cdot d\mathbf{a}.$$

- $[\psi_\nu; \psi_\mu] = 0$, if ν and μ relate to different ports.
- If ν and μ relate to the same port plane p :

$$[\psi_\nu^r; \psi_\mu^d] = \iint_p (E_{\nu x}^r H_{\mu y}^d - E_{\nu y}^r H_{\mu x}^d - H_{\nu y}^r E_{\mu x}^d + H_{\nu x}^r E_{\mu y}^d) dx_p dy_p.$$

Scattering matrices, reciprocity

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- Compare with the modal orthogonality relations on port plane p ,

$$(\psi_\nu^r; \psi_\mu^d) = \frac{1}{4} \iint_p ((E_{\nu x}^r)^* H_{\mu y}^d - (E_{\nu y}^r)^* H_{\mu x}^d + (H_{\nu y}^r)^* E_{\mu x}^d - (H_{\nu x}^r)^* E_{\mu y}^d) dx_p dy_p.$$

Scattering matrices, reciprocity

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- Compare with the modal orthogonality relations on port plane p , for propagating modes with real transverse components:

$$(\psi_\nu^r; \psi_\mu^d) = \frac{1}{4} \iint_p (E_{\nu x}^r H_{\mu y}^d - E_{\nu y}^r H_{\mu x}^d + H_{\nu y}^r E_{\mu x}^d - H_{\nu x}^r E_{\mu y}^d) dx_p dy_p,$$

$$(\psi_\nu^f; \psi_\mu^f) = \delta_{\nu\mu} P_\nu, \quad (\psi_\nu^b; \psi_\mu^b) = -\delta_{\nu\mu} P_\nu, \quad (\psi_\nu^f; \psi_\mu^b) = (\psi_\nu^b; \psi_\mu^f) = 0.$$

Scattering matrices, reciprocity

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- $\psi^f = (E_x, E_y, iE_z, H_x, H_y, iH_z)^\top$
 $\psi^b = (E_x, E_y, -iE_z, -H_x, -H_y, iH_z)^\top$. (Real components).

Scattering matrices, reciprocity

- If ν and μ relate to the same port plane p :

$$[\psi_\nu^r; \psi_\mu^d] = \iint_p (E_{\nu x}^r H_{\mu y}^d - E_{\nu y}^r H_{\mu x}^d - H_{\nu y}^r E_{\mu x}^d + H_{\nu x}^r E_{\mu y}^d) dx_p dy_p.$$

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 $\psi^b = (E_x, E_y, -iE_z, -H_x, -H_y, iH_z)^\top$. (Real components).

$$[\psi_\nu^f; \psi_\mu^f] = [\psi_\nu^b; \psi_\mu^b] = 0, \quad [\psi_\nu^f; \psi_\mu^b] = -\delta_{\nu\mu} 4P_\nu, \quad [\psi_\nu^b; \psi_\mu^f] = \delta_{\nu\mu} 4P_\nu.$$

Scattering matrices, reciprocity

$$\hookrightarrow 0 = \sum_{\nu} 4P_{\nu} (B_{1,\nu} F_{2,\nu} - F_{1,\nu} B_{2,\nu}),$$

uniform normalization $P_{\nu} = P_0$,

$$\hookrightarrow 0 = \sum_{\nu} (B_{1,\nu} F_{2,\nu} - F_{1,\nu} B_{2,\nu}),$$

$$\hookrightarrow 0 = \mathbf{B}_1 \cdot \mathbf{F}_2 - \mathbf{F}_1 \cdot \mathbf{B}_2,$$

$$\mathbf{F}_j = \mathbf{S} \mathbf{B}_j,$$

$$\hookrightarrow 0 = \mathbf{B}_1 \cdot \mathbf{S} \mathbf{B}_2 - (\mathbf{S} \mathbf{B}_1) \cdot \mathbf{B}_2,$$

$$\hookrightarrow 0 = \mathbf{B}_1 \cdot \mathbf{S} \mathbf{B}_2 - \mathbf{B}_1 \cdot \mathbf{S}^{\top} \mathbf{B}_2,$$

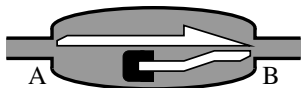
$$\hookrightarrow 0 = \mathbf{B}_1 \cdot (\mathbf{S} - \mathbf{S}^{\top}) \mathbf{B}_2 \text{ for all } \mathbf{B}_1, \mathbf{B}_2.$$

$$\mathbf{S} = \mathbf{S}^{\top}, \quad S_{\nu\mu} = S_{\mu\nu} \text{ for all } \nu, \mu.$$

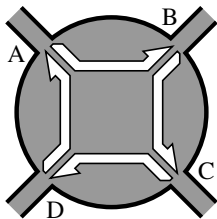
The scattering matrix of a *reciprocal circuit* is *symmetric*.

Reciprocal circuit: made of reciprocal media, with $\hat{\epsilon} = \hat{\epsilon}^{\top}$, $\hat{\mu} = \hat{\mu}^{\top}$.

Nonreciprocal devices



Isolator:
unidirectional transmission,
 $S_{BA} = 1, S_{AB} = 0.$



Circulator:
transmission cycle,
 $S_{BA} = 1, S_{CB} = 1, S_{DC} = 1, S_{AD} = 1,$
 $S_{..} = 0$ otherwise.

Required: nonreciprocal media with $\hat{\epsilon} \neq \hat{\epsilon}^T$,
↔ magneto optic media, Faraday effect.

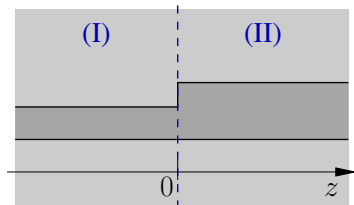
Nonreciprocal devices

What about, for example,

- a long, “adiabatic” Y-junction ?
- a junction between a single mode core and a wider multimode waveguide ?

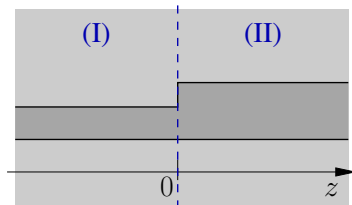


Waveguide discontinuities



Half-infinite waveguides (I), (II),
discontinuity at $z = 0$.

Waveguide discontinuities



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discontinuity at $z = 0$.

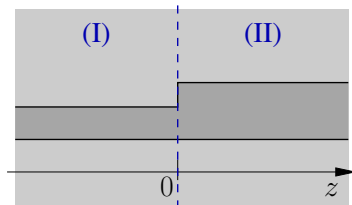
- Expand into local normal modes
 $\{\psi_{s,m}^d, \beta_{s,m}\}$, $m \in \mathcal{N}_s$, $s = \text{I, II}$:
 Transverse boundary conditions \leftrightarrow discrete sets.

$$\begin{pmatrix} \mathbf{E} \\ \mathbf{H} \end{pmatrix}_s(x, y, z) = \sum_{m \in \mathcal{N}_s} \left\{ f_{s,m} \psi_{s,m}^f(x, y) e^{-i\beta_{s,m}z} + b_{s,m} \psi_{s,m}^b(x, y) e^{+i\beta_{s,m}z} \right\},$$

$z < 0$: $s = \text{I}$, $f_{\text{I},m}$ given influx, $b_{\text{I},m}$ unknown,
 $z > 0$: $s = \text{II}$, $f_{\text{II},m}$ unknown, $b_{\text{II},m}$ given influx.

$\hookrightarrow (\mathbf{E}, \mathbf{H})_{\text{I,II}}$ are solutions for $z < 0$ and $z > 0$.

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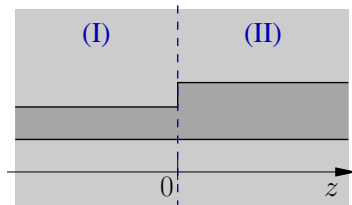
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- Continuity of the tangential components of \mathbf{E}, \mathbf{H} at the interface
 \leftrightarrow formally equate expressions for $(\mathbf{E}, \mathbf{H})_{\text{I,II}}$ at $z = 0$.

(Only equality of E_x, E_y, H_x, H_y will be relevant.)

Waveguide discontinuities



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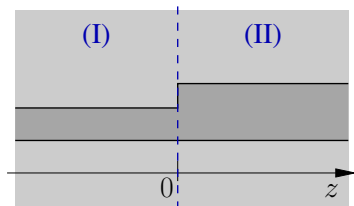
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- Project on $\psi_{s,l}^d$ to extract coefficients ...

Waveguide discontinuities, scattering matrix



(Global coordinate $z \neq$ former local coordinate on port I)
(One variant of a projection procedure.)

- $(\psi_{I,l}^b; \cdot = \cdot)$, $l \in \mathcal{N}_I$:

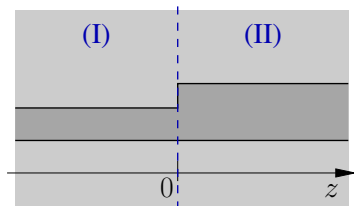
$$\sum_{m \in \mathcal{N}_I} [f_{I,m}(\psi_{I,l}^b; \psi_{I,m}^f) + b_{I,m}(\psi_{I,l}^b; \psi_{I,m}^b)] = \sum_{m \in \mathcal{N}_{II}} [f_{II,m}(\psi_{I,l}^b; \psi_{II,m}^f) + b_{II,m}(\psi_{I,l}^b; \psi_{II,m}^b)],$$

- $(\psi_{II,l}^f; \cdot = \cdot)$, $l \in \mathcal{N}_{II}$:

$$\sum_{m \in \mathcal{N}_I} [f_{I,m}(\psi_{II,l}^f; \psi_{I,m}^f) + b_{I,m}(\psi_{II,l}^f; \psi_{I,m}^b)] = \sum_{m \in \mathcal{N}_{II}} [f_{II,m}(\psi_{II,l}^f; \psi_{II,m}^f) + b_{II,m}(\psi_{II,l}^f; \psi_{II,m}^b)],$$

$$\left(\begin{array}{c} \mathbf{b}_I \\ \mathbf{f}_I \end{array} \right) = \mathbf{S} \left(\begin{array}{c} \mathbf{f}_I \\ \mathbf{b}_{II} \end{array} \right) = \left(\begin{array}{cc} \mathbf{S}_{I,I} & \mathbf{S}_{I,II} \\ \mathbf{S}_{II,I} & \mathbf{S}_{II,II} \end{array} \right) \left(\begin{array}{c} \mathbf{f}_I \\ \mathbf{b}_{II} \end{array} \right).$$

Waveguide discontinuities, overlap model



Most simplified variant:
Unidirectional **overlap model**.

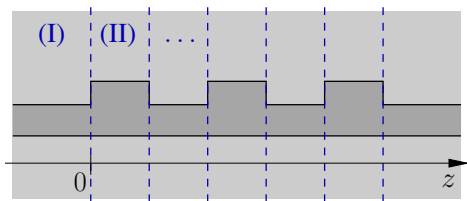
- (I): Incoming guided mode ψ_I , reflections & radiation neglected.
(II): Outgoing guided modes $\psi_{II,m}$, radiation neglected.

- $f_I \psi_I \approx \sum_m f_{II,m} \psi_{II,m}$ at $z = 0$.

↪
$$f_{II,m} = \frac{(\psi_{II,m}; \psi_I)}{(\psi_{II,m}; \psi_{II,m})} f_I, \quad \text{or} \quad f_{II,m} = \frac{1}{P_{II,m}} (\psi_{II,m}; \psi_I) f_I.$$

(Transmission is given directly by the “overlaps” ↪ Relevance of the mode products $(\cdot; \cdot)$.
(Cf. explicit expressions for overlaps of 2-D modes, involving only principal mode profile components.)

A sequence of waveguide discontinuities



- Divide into segments.
- Establish local normal mode expansions.
- Project on local modes.

↪ Linear system of equations for all local mode amplitudes.

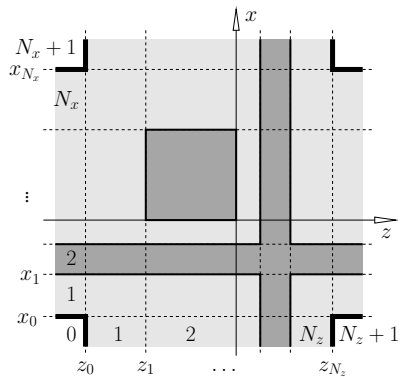
↪ Solve (...) \rightsquigarrow $\begin{pmatrix} E \\ H \end{pmatrix} (x, y, z)$.

Bidirectional eigenmode propagation (BEP),
Eigenmode expansion method (EME),

...

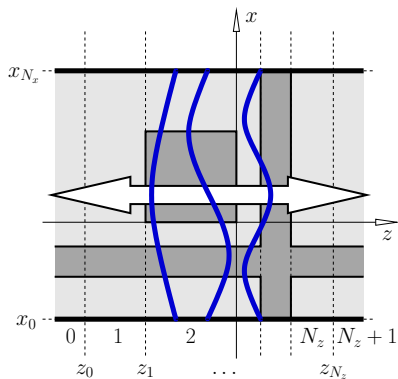
(Radiated outgoing fields: Open boundary conditions required (PMLs) \leftrightarrow Complex eigenmodes.)
(2-D: ok. 3-D: ?)

Rectangular 2-D circuits



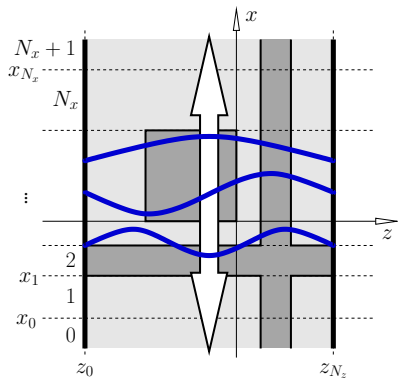
- Divide into slices & layers.

Rectangular 2-D circuits



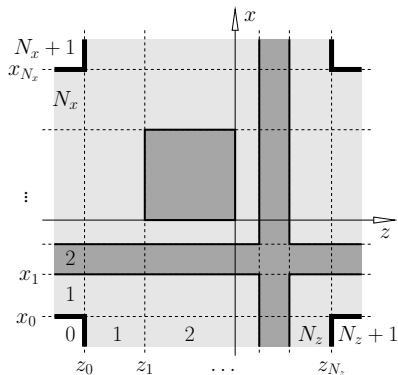
- Divide into slices & layers.
- Establish local modes:
Propagation along $\pm z$,
boundary conditions $\phi = 0$.

Rectangular 2-D circuits



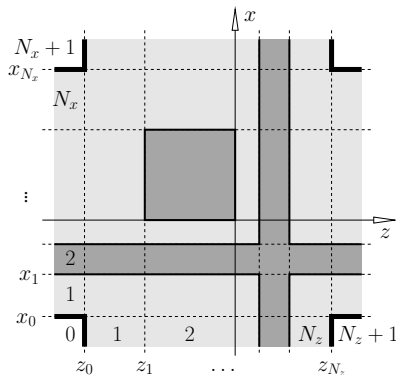
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Rectangular 2-D circuits

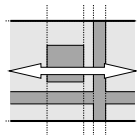


- Divide into slices & layers.
- Establish local modes:
 - Propagation along $\pm z$,
 - & Propagation along $\pm x$,
 - boundary conditions $\phi = 0$.
- Project at horizontal & vertical interfaces.

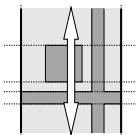
Rectangular 2-D circuits



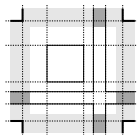
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 & Propagation along $\pm x$,
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horizontal BEP,

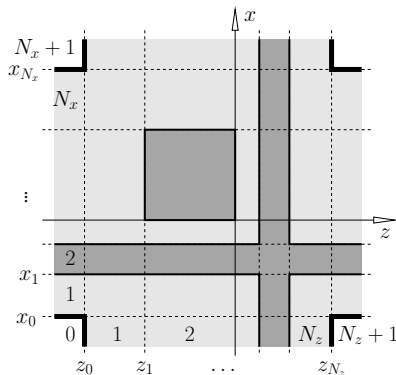


vertical BEP,



continuity at x_0, x_N, z_0, z_N .

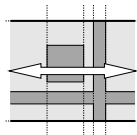
Rectangular 2-D circuits



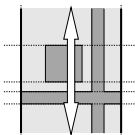
Quadridirectional Eigenmode Propagation (QUEP)



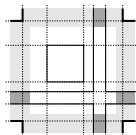
- Divide into slices & layers.
- Establish local modes:
 Propagation along $\pm z$,
 & Propagation along $\pm x$,
 boundary conditions $\phi = 0$.
- Project at horizontal
 & vertical interfaces.



horizontal BEP,



vertical BEP,



continuity at $x_0, x_{N_x}, z_0, z_{N_z}$.

Upcoming

Next lectures:

- Bent optical waveguides; whispering gallery resonances; circular microresonators.
- Coupled mode theory, perturbation theory.
- A touch of photonic crystals; a touch of plasmonics.

